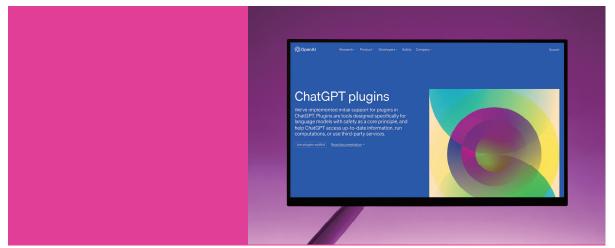
DISCRETE-TIME LINEAR TIME-INVARIANT SYSTEMS CHAPTER 10



10.1 IMPULSE REPRESENTATION OF DISCRETE-TIME SIGNALS

IMPULSE REPRESENTATION OF DISCRETE-TIME SIGNALS

$$\delta[n - n_0] = \begin{cases} 1, & n = n_0 \\ 0, & n \neq n_0 \end{cases}$$

$$x[n]\delta[n-n_0] = x[n_0]\delta[n-n_0].$$

• We define the signals

$$\begin{aligned} x_{-1}[n] &= x[n]\delta[n+1] = x[-1]\delta[n+1] = \begin{cases} x[-1], & n = -1\\ 0, & n \neq -1 \end{cases} \\ x_0[n] &= x[n]\delta[n] = x[0]\delta[n] = \begin{cases} x[0], & n = 0\\ 0, & n \neq 0 \end{cases} \\ x_1[n] &= x[n]\delta[n-1] = x[1]\delta[n-1] = \begin{cases} x[1], & n = 1\\ 0, & n \neq 1 \end{cases} \end{aligned}$$

10.2 Convolution for discrete-time systems

Property of convolution sum:

$$\delta[n] * g[n - n_0] = \delta[n - n_0] * g[n] = g[n - n_0]$$

$$\delta[n]g[n - n_0] = g[-n_0]\delta[n]$$

$$\delta[n - n_0]g[n] = g[n_0]\delta[n - n_0],$$

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n - k].$$

Representation of a signal with discrete-time impulse functions.

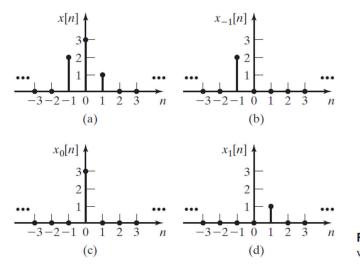


Figure 10.1 Representation of a signal with discrete-time impulse functions.

Convolution sum for DT signals

h[n]

 $\delta[n]$ System

Figure 10.2 Impulse response of a system.

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] = x[n]*h[n]$$
$$= \sum_{k=-\infty}^{\infty} h[k]x[n-k] = h[n]*x[n].$$

To illustrate a property of the convolution sum, we calculate the output at n = 0.

$$y[0] = \cdots + x[-2]h[2] + x[-1]h[1] + x[0]h[0] + x[1]h[-1] + x[2]h[-2] + \cdots.$$

EXAMPLE 10.1 A Finite Impulse Response System

• The system difference equation directly from Figure 10.3 is

$$y[n] = (x[n] + x[n - 1] + x[n - 2])/3.$$

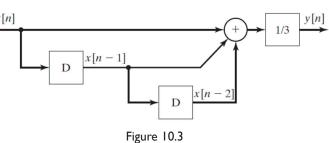
$$h[n] = (\delta[n] + \delta[n - 1] + \delta[n - 2])/3|_{n=0} = (1 + 0 + 0)/3 = 1/3;$$

$$h[1] = (\delta[n] + \delta[n - 1] + \delta[n - 2])/3|_{n=1} = (0 + 1 + 0)/3 = 1/3;$$

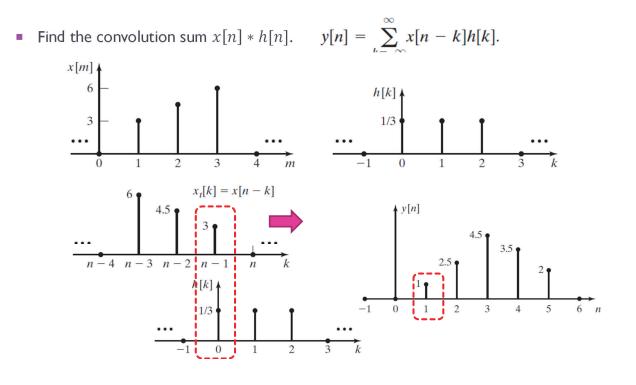
$$h[2] = (\delta[n] + \delta[n - 1] + \delta[n - 2])/3|_{n=2} = (0 + 0 + 1)/3 = 1/3;$$

$$h[n] = 0, \text{ all other } n.$$
is a finite impulse response (FIR)
$$x[n] = \frac{x[n]}{1/3}$$

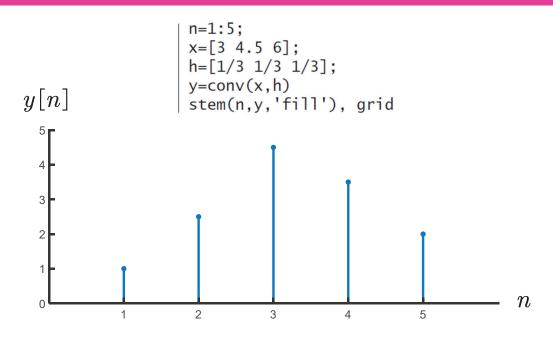
This is a finite impulse response (FIR) system; that is, the impulse response contains a finite number of nonzero terms.



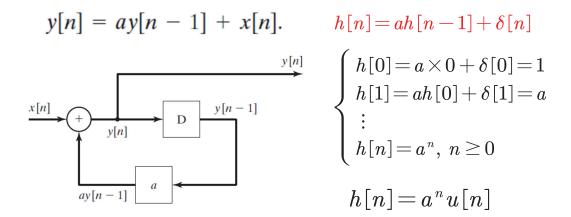
EXAMPLE 10.2 (必考題) System Response by Convolution for An LTI System



MATLAB® code of convolution sum



EXAMPLE 10.3 Calculation of the impulse response of a discrete system



The unit impulse response consists of an unbounded number of terms; this system is called an infinite impulse response (IIR) system

EXAMPLE 10.4 Step response of a discrete system with $h[n] = 0.6^n u[n]$

• Find the unit step response of this system, with x[n] = u[n].

-1

0

1

3

4

5

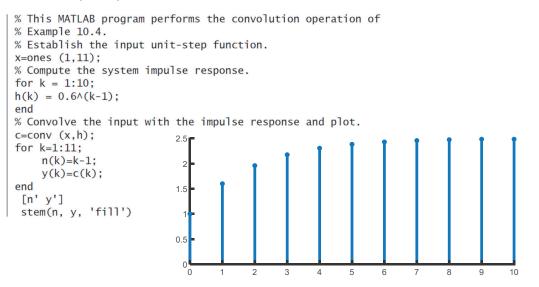
n

$$y[n] = \sum_{k=-\infty}^{\infty} x[n-k]h[k] = \sum_{k=-\infty}^{\infty} u[n-k](0.6)^{k}u[k] = \sum_{k=0}^{n} (0.6)^{k},$$

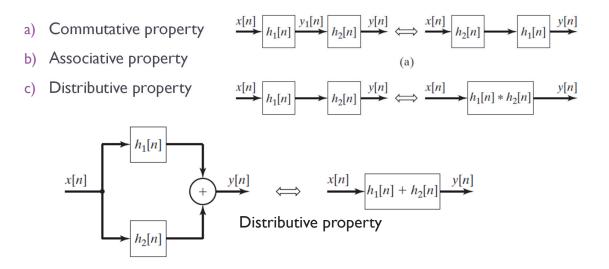
$$\sum_{k=0}^{n} a^{k} = \frac{1-a^{n+1}}{1-a}.$$
$$y[n] = \sum_{k=0}^{n} (0.6)^{k} = \frac{1-(0.6)^{n+1}}{1-0.6} = 2.5[1-(0.6)^{n+1}], \quad n \ge 0.$$

MATLAB® EXAMPLE CODE

Unit step response



Properties of convolution Sum



10.3 Properties of Discrete-time LTI Systems

- Memory
 - A memoryless LTI system is then a pure gain, described by y[n] = Kx[n].
- Invertibility

 $h[n]*h_i[n] = \delta[n],$

- We do not present a procedure for finding the impulse response given h[n]. This problem can be solved with the use of the *z*-transform of Chapter 11.
- Causality

$$y[n_1] = T(x[n]), \quad n \leq n_1,$$

- As an additional point, a signal that is zero for is called a *causal signal*
- For a causal LTI discrete-time system, h[n] = 0 for t < 0.

10.3 Properties of Discrete-time LTI Systems

- Stability
 - h[n] is said to be absolutely summable. $\iff \sum_{k=-\infty}^{\infty} |h[k]| < \infty$.

EXAMPLE 10.6

Stability of an LTI discrete system with $h[n] = \left(\frac{1}{2}\right)^n u[n]$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=0}^{\infty} \left(\frac{1}{2}\right)^n = \frac{1}{1-1/2} = 2.$$

10.4 Difference-Equation Models

A first-order discrete-time system model

y[n] = ay[n-1] + bx[n].

 The general form of an Nth-order linear difference equation with constant coefficients is,

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k], \quad a_0 \neq 0.$$

It can be shown by these procedures that this equation is both linear and time invariant.

EXAMPLE 10.11 Stability of a discrete system

Suppose that a causal system is described by the difference equation

$$y[n] - 1.25y[n - 1] + 0.375y[n - 2] = x[n].$$

特性方程式 $1-1.25z^{-1}+0.375z^{-2}=0$,

 $z^2 - 1.25z + 0.375 = 0, \rightarrow z = 0.75, 0.5$

The natural response is given by

 $y_c[n] = C_1(0.75)^n + C_2(0.5)^n.$

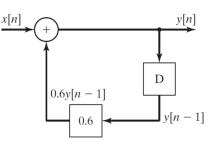
This function approaches zero as n approaches infinity. It's a stable system.

All roots satisfy |z| < 1, then the system is stable.

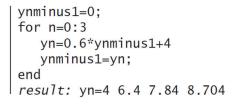
EXAMPLE 10.12

Simulation diagram for a discrete system

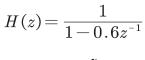
Considered a discrete-time system described by the difference equation y[n] - 0.6y[n - 1] = x[n].



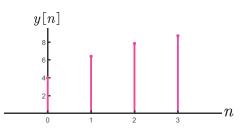
MATLAB program with x[n] = 4.



The transfer function

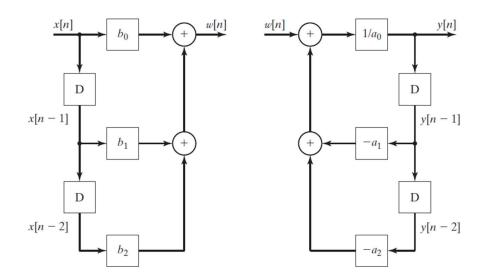


$$=\frac{z}{z-0.6}$$



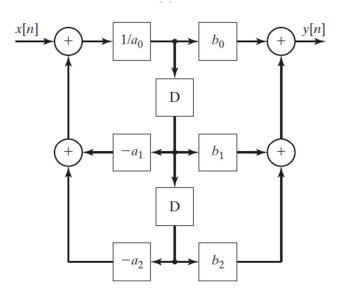
BLOCK DIAGRAM (TYPE I)

 $a_0y[n] + a_1y[n-1] + a_2y[n-2] = b_0x[n] + b_1x[n-1] + b_2x[n-2].$



BLOCK DIAGRAM (TYPE II)

 $a_0y[n] + a_1y[n-1] + a_2y[n-2] = b_0x[n] + b_1x[n-1] + b_2x[n-2].$



EXAMPLE 10.14 Transfer function for a discrete system

• The α -filter

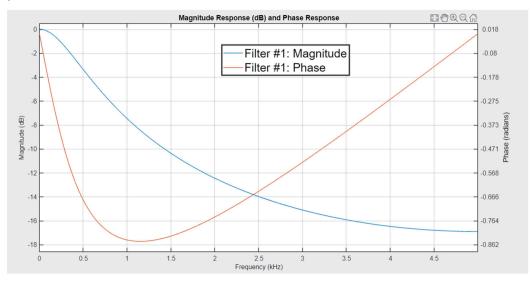
$$y[n] - (1 - \alpha)y[n - 1] = \alpha x[n].$$

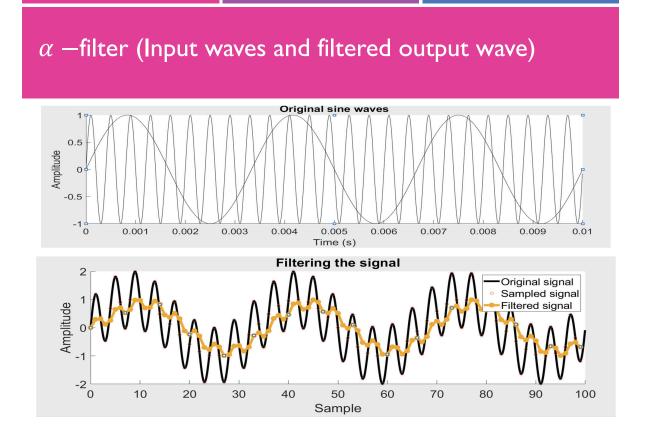
$$H(z) = \frac{\alpha}{1 - (1 - \alpha)z^{-1}} = \frac{\alpha}{z - (1 - \alpha)}.$$

$$\xrightarrow{x[n]} \boxed{\frac{\alpha z}{z - (1 - \alpha)}} \xrightarrow{y[n]}$$
Figure 10.21 α -Filter.

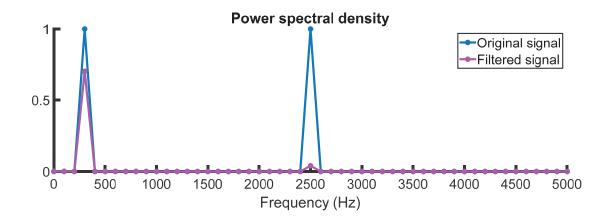
Frequency response of an α –filter

Alpha-filter





Spectra of Input and Filtered Waves of the α -filter



LTI System Input-Output Functions

LTI systems

TABLE 10.1 Input–Output Functions for an LTI System

$$H(z) = \sum_{k=-\infty}^{\infty} h[k] z^{-k}$$
$$X z_1^n \to X H(z_1) z_1^n, \quad X = |X| e^{j\phi}$$
$$|X| \cos(\Omega_1 n + \phi) \to |X| |H(e^{j\Omega_1})| \cos(\Omega_1 n + \phi + \theta_H)$$

 $\begin{array}{c} x[n] \\ & h[n] \\ & & \downarrow \\ & h[n] \\ & & \downarrow \\ & H[z] = \sum_{k=-\infty}^{\infty} h[k] z^{-k} \end{array}$

Figure 10.22 LTI system.



CHAPTER 10

- THE END -